

#### Actuator Design for a Passive Haptic Display

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Research Supported by National Science Foundation Grant IIS-9700528



Overview

- Background
- Description of existing actuators and their deficiencies
- Alternative clutches / brakes
- Redesign of PTER's existing Dynacorp brakes / clutches
- Calibration of the new torque sensor
- Digital feedback control, experimental & simulation
- Conclusion
- Future work
- Questions



## Background



Passive Trajectory Enhancing Robot (PTER)

Virtual Corridor simulated on PTER

- PTER uses friction brakes to constrict motion for simulation of haptic features.
- Over actuated



## PTER's Existing Brakes / Clutches

- Industrial electromagnetic friction units from Dynacorp
  - Model 310
  - Original max torque
    300 ft-lbf (407 N-M)
  - Rated time constant
    0.105 sec (coil build up)
- Modified to eliminate metal to metal contact
  - Reduced available torque
- No provisions of measuring actual applied torque for feedback control







## Dynamic Response of PTER's Actuators

- Open Loop Control
- Undesirable Dynamics
  - Non-linear electromagnet
  - Sliding on pins (binding)
  - First order response
    - R-L circuit
  - Pure time delay
    - Coil build up to attract armature plate
  - Steady state error
    - Each clutch's output torque different
    - Max torque ranged from 15 to 125 Ft-Lbf, depending on unit



(Borrowed from Gomes, 1997)



## Alternative Brakes / Clutches

- Hysteresis
  - Low torque rattings (2.5 oz-in (0.013 lbf-ft) -- 3200 oz-in (16.67 lbf-ft))
  - Cogging
  - Hysteresis in actuation curve
- Eddy Current
  - Torque dependent on both slip velocity and applied current
  - Large scale applications
- Magnetic Particle
  - Required to turn a full revolution for redistribution of magnetic particles (for consistent operation)
- Electro-Rheological
  - Not much information available



# Electromagnetic Friction

- Metal to metal contact for completing the magnetic circuit
- Friction material is only used for stabilizing torque
- Max torque
  - 5 oz-in (0.026 lbf-ft) to 465 lbf-ft
- Response consists of both mechanical and electrical (applied voltage)
- Dynamic times depend on size of the unit (Kebco)
  - Pure time delay:  $t_{11} = 0.004 0.055$  sec.
  - Rise time:  $t_{12} = 0.006 0.240$  sec.
  - Engagement time:  $t_1 = 0.01 0.295$  sec.





# New Design Concept With Torque Sensor

- Spoke transmits torque from friction
- Spoke locates and supports armature plate
- Spoke deflects under vertical engagement and torque transmission
- Strain gauge measurement proportional to transmitted torque





#### Modified Clutch Layout (Top View)





#### Modified Clutch (Section View)





## Spoke Def - Bod Model





## Integration Constants (Torque Bending)

[	0	R	- 1	0	0	0	0				[ 0 ]	
	$\frac{-1}{2} \cdot \frac{\left(2 \cdot a1 \cdot R + a1^2\right)}{(E2 \cdot I2)}$	$\frac{1}{(E2\cdot I2)}$	0	-1 (E3·I3)	0	0	0				- a1 · T E2 · I2	
	$\frac{-1}{6} \cdot \frac{\left(3 \cdot a1^2 \cdot R + a1^3\right)}{(E2 \cdot I2)}$	$\frac{\text{al}}{(\text{E2-I2})}$	1 (E2·I2)	0	-1 (E3·I3)	0	0		F C2		$\frac{-1}{2} \cdot \frac{a l^2 \cdot T}{(E2 \cdot I2)}$	
	$\frac{-1}{2} \cdot \frac{\left(2 \cdot b \cdot R + 2 \cdot b \cdot a1 + b^2\right)}{(E3 \cdot I3)}$	0	0	1 (E3·I3)	0	-1 (E4·I4)	0	x	D2 C3	=	- b·T E3·I3	
	$\frac{-1}{6} \cdot \frac{\left< 3 \cdot b^2 \cdot a1 + b^3 + 3 \cdot b^2 \cdot R \right>}{(E3 \cdot I3)} =$	0	0	<mark>ь</mark> (ЕЗ-ІЗ)	1 (E3·I3)	0	-1 (E4·I4)		D3 C4		$\frac{-1}{2} \cdot \frac{b^2 \cdot T}{(E3 \cdot I3)}$	
	$\frac{-1}{2} \cdot \frac{\left\langle 2 \cdot a 2 \cdot L + 2 \cdot a 2 \cdot R - a 2^2 \right\rangle}{(E4 \cdot I4)} =$	0	0	0	0	$\frac{1}{(E4\cdot I4)}$	0		[D4]		$\frac{-a2 \cdot T}{E4 \cdot I4}$	
-	$\frac{1}{6} \cdot \frac{\left(3 \cdot a2^2 \cdot R - 2 \cdot a2^3 + 3 \cdot a2^2 \cdot L\right)}{(E4 \cdot I4)}$	0	0	0	0	a2 (E4·I4)	1 (E4·I4)				$\left[\frac{-1}{2} \cdot \frac{a2^2 \cdot T}{(E4 \cdot I4)}\right]$	

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## Electromagnet Model



Force \* 
$$\delta X = \delta Energy$$
  
Force =  $\frac{d(E)}{dx} = -\frac{1}{2} \frac{L_0}{a} \frac{1}{(1+x/a)^2} i^2$ 

Using Principle of Virtual Work a≈0.0113, L<sub>0</sub>≈11.298



## **Design Considerations**

- Available magnetic force
- Cyclic fatigue
- Physical size constraints
- Sensor sensitivity
- Material selection
  - Spokes
    - Delrin 100P
    - Steel ASTM-A514
    - Aluminum 7075-T651
  - Armature Plate
    - Low carbon steel (under 1010)
- Alternative friction material







# Torque Bar Calibration



- Strain measured with strain gauges in a half bridge
- Simple beam theory used to relate applied torque
- Spring scale used to apply approximate known torque
- Measurements compared, found to be sufficiently close



# Clutch's Sensor Calibration

- Strain in spokes measured by strain gauges in a full bridge
- First order equation fitting strain to measured torque from torque bar
- Model predicted a scaling factor of  $3.361x10^5$ 
  - 9.94% discrepancy from actual
- Model's predicted spring constants
  - Axial --  $1.77x10^3$  lbf/in
  - Radial -- 2,217 lbf-in/deg



$$\tau = 3.057 x 10^5 \varepsilon_{spoke} + 4.818 \text{ (in - lbf)}$$



#### Torque vs. Current Calibration





## **Digital Control**



$$\tau = ai^{2} + bi + c \Longrightarrow k_{clutch} = \frac{d\tau}{di} = 2ai + b \qquad \tau = \tau_{eq} + \frac{k_{p}k_{clutch}}{1 + k_{p}k_{clutch}}(\tau_{d} - \tau_{eq}) + \frac{1}{1 + k_{p}k_{clutch}}\tau_{dist}$$

- Nonlinear feed-forward based on quadratic torque mapping
- Absolute value of torque fed back (direction insensitive)
- Proportional control based on error
- Larger gain equates to larger disturbance and error rejection



## **Experimental Results**





- Labview processing controller at a non-deterministic 50 Hz  $(t_s \approx 0.02 \text{ sec})$
- Small gains stabilized torque
- Large gains caused system to go unstable



Power Supply Model



- Second order response
  - t<sub>r</sub>  $\approx 0.056$  sec
  - -%OS  $\approx 25\%$

$$PS(s) = \frac{I(s)}{I_c(s)} = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2} = \frac{3751}{s^2 + 49s + 3751}$$

- $-\omega_n = 61.25 \text{ rad/sec}$
- $\xi = 0.40$
- Voltage applied to a first order system (RL Circuit)
  - $R \approx 11.68 \Omega$
  - $-L \approx 1.19H$
- Voltage saturation: 0 36 volts



## Power Supply Data vs. Simulation





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# Digital Stability -- Gain Margin

- $i \approx 0.9$  amps for unstable experiment
  - k<sub>clutch</sub> = 435 in-lbf / amp



- Gain margin = 5.011 db
- Max system gain:  $k_p k_{clutch} = 1.78$
- Max controller gain:  $k_p = 0.0041$  amps / in-lbf
- Error & disturbance rejection to 36%
- ts = 0.002 (500 Hz)  $\longrightarrow$   $Ps(z) = \frac{0.007254z + 0.007021}{z^2 1.892z + 0.9066}$ 
  - Gain margin = 22.475 db
  - Max system gain:  $k_p k_{clutch} = 13.30$
  - Max controller gain:  $k_p = 0.0306$  amps / in-lbf
  - Error & disturbance rejection to 7%



# Simulations for Regulating @ 250 in-lbf



- Increased sampling rate allows increased proportional gain • and increased error rejection
- Strong evidence that instability is caused by relatively slow controller and quick second order underdamped power supply dynamics 23



#### **Digital PD Control**





# Digital PD Regulating @ 250 in-lbf



(Simulations)



# Reference Tracking (Digital PD)



- $\alpha = 0.5$ 
  - $k_{p}k_{clutch} = 8.125$  $- k_{d}k_{clutch} = 8.125$
- Online gain normalization based on estimated clutch gain from feed-forward actuation.

(Simulations)



## Stick / Slip of the Armature Plate

- Determine if the armature plate comes to rest, sticking then slipping, causing the instability or limit cycle
- Compare angular velocity of the Shaft hub with the angular velocity of the armature plate
- Numerically differentiate and filter data
- Armature plate appears to not come to rest while hub is moving



$$\tau = k(\theta_1 - \theta_2) \Longrightarrow \dot{\tau} = k(\dot{\theta_1} - \dot{\theta_2})$$

(Data from unstable experiment displayed in slide 19) 27



# **Closing Remarks -- Conclusion**

- Industrial available clutches / brakes were not found appropriate for our application.
- Redesigned electromagnetic clutch incorporating a torque sensor for feedback control was developed and one prototype was built.
- More thorough modeling & testing of the new clutch and power supply combination.
- Feasibility and benefits of closed loop control were demonstrated.
- Power Supply dynamics became a dominant factor.
- Faster digital sampling required for effective disturbance rejection



# Closing Remarks -- Future Work -- I

- Faster Digital Controller
  - dSPACE & Real Time workshop
  - Hyperkernal
- Online Parameter Identification
- Analog Control
- Alternative Friction Materials
  - Delrin:  $\mu_s = 0.20$ ,  $\mu_d = 0.35$
- Power Supply Upgrade
  - Bipolar
  - Larger current and voltage ratings
- Design Modifications



## Closing Remarks -- Future Work -- II

- Manufacture three more units
- Upgrade other components of PTER
  - Replace position potentiometers (Encoders, Resolvers)
  - Tachometers
  - Remount strain gauges in force sensors
  - Computer and Software
- Programming of old and new haptic algorithms
- Bilateral control of Hurbirt
- Possible improvements to the design of PTER



Questions

- What do you mean?
- How can you justify that?
- I don't agree!
- Why didn't you....?
- Did you try....?
- Etc?



# Digital Control of a fast Power Supply

$$PS(z) = Z[Zoh * PS(s)] = Z\left[\frac{1 - e^{-Ts}}{s} * \frac{1}{\frac{s}{a} + 1}\right] = (1 - z^{-1}) * Z\left[\frac{1}{s(\frac{s}{a} + 1)}\right] = \frac{1 - e^{-aT}}{z - e^{-aT}}$$

- With an open loop gain of "1", disturbance rejection is limited to only 50%
- Systems dynamics become much quicker then the digital controller



# Tuning of the RL circuit

- Pro
  - Addition of resistors and capacitors can tune the circuit for desirable second order dynamics
- Con
  - Limited by achievable output
- Increasing "R" increases damping and required voltage
- Large "L" causes low damping and low natural frequency
- Small "C" increases natural frequency.





$$\frac{I_{clutch}(s)}{I_{supply}(s)} = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2} = \frac{\frac{1}{CL}}{s^2 + \frac{R}{L}s + \frac{1}{CL}}$$



# Stick / Slip Dynamic Model

$$\begin{bmatrix} I_{shaft+Hub} & 0\\ 0 & I_{Armature} \end{bmatrix} \begin{bmatrix} \ddot{\theta}_{shaft+Hub}\\ \ddot{\theta}_{Armature} \end{bmatrix} + \begin{bmatrix} B_1 + B_2 & -B_2\\ -B_2 & B_2 \end{bmatrix} \begin{bmatrix} \dot{\theta}_{shaft+Hub}\\ \dot{\theta}_{Armature} \end{bmatrix} + \begin{bmatrix} k & -k\\ -k & k \end{bmatrix} \begin{bmatrix} \theta_{shaft+Hub}\\ \theta_{Armature} \end{bmatrix} = \begin{bmatrix} \tau_{in}\\ \tau_{friction} \end{bmatrix}$$

- Account for braking force through the armature plate, compliant spoke, and to the hub.
- Karnopp model utilized for friction force.
- Damping added to account for viscous friction in the bearings & material damping in the spoke
- Stiff spoke and small armature plate results in a stiff system that Simulink can not successfully integrate.
- Fixed step simulations gave erroneous information





PTER's original clutch/brake.



Section view of clutch/brake.

Industrial electromagnetic friction clutch / brakes (Dynacorp Model 310).

- Time constant for coil build up:  $\tau = 0.105$ sec (given applied voltage)
- Original max rated torque = 300 Ft-Lbf (407 N-M)
- Max voltage: V = 24 volts
- Max current: I = 2.376 amps
- Non asbestos friction material with coefficient of friction:  $\mu = 0.45$
- Modified to eliminate metal to metal contact (reduce max torque capabilities)



# Hysteresis & Eddy Current

Hysteresis

- Max torque ratings
  - 2.5 oz-in (0.013 lbf-ft) -- 3200 oz-in (16.67 lbf-ft)
- No rubbing parts
- Hysteresis in the torque vs current mapping
- Cogging

#### Eddy Current

- Torque dependent on both slip velocity and applied current
- Large scale applications







## Magnetic Particle

- Max torque
  - 5 oz-in (0.026 lbf-ft) to 738 lbf-ft
- Torque generated by shear and tensile stress between attracted magnetic particles
- Dynamic time constants ranging from 0.009 to 1.31 sec., depending on the size of the unit.
- Clean sealed operation
- Required to turn a full revolution to redistribute magnetic particles (for consistent operation)

